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Statistical characterization of Time-Dependent Variability defects using the Maximum Current Fluctuation

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Abstract— This paper presents a new methodology to extract, at a given operation condition, the statistical distribution of the number of active defects that contribute to the observed device Time-Dependent Variability, as well as their amplitude distribution. Unlike traditional approaches based on complex and time-consuming individual analysis of thousands of current traces, the proposed approach uses a simpler trace processing, since only the maximum and minimum values of the drain current during a given time interval are needed. Moreover, this extraction method can also estimate defects causing small current shifts, which can be very complex to identify by traditional means. Experimental data in a wide range of gate voltages, from near-threshold up to nominal operation conditions, are analyzed with the proposed methodology.

Index Terms—Time-Dependent Variability (TDV), Random Telegraph Noise (RTN), Bias Temperature Instability (BTI), transistor, Maximum Current Fluctuation (MCF)

I. INTRODUCTION

TIME-Dependent Variability (TDV), caused by phenomena such as Random Telegraph Noise (RTN) and Bias Temperature Instability (BTI) has become a subject of increasing concern in deeply-scaled CMOS technologies [1], due to its role as a source of device and circuit performance variability [2], [3]. At device level, RTN is observed as discrete jumps of the drain current. It is generally accepted that these current shifts are caused by the stochastic trapping/detrapping of charge carriers in/from defects. In particular, when charge carriers are trapped in defects, they change the density of charge located at the silicon-oxide interface, which causes an increase in the threshold voltage of the device, thus decreasing its drain current [4],[5]. Detrapping of carriers from defects has therefore the opposite impact on the device current. Because of the stochastic nature of the phenomenon, the threshold voltage shifts (or, equivalently, current jumps), are statistically

distributed. BTI is also associated to the same kind of phenomena, though, in this case, charge de-trapping is mainly observed when gate bias is decreased (a phase known as recovery). However, some of the carriers may not be de-trapped, leading to a permanent shift of the threshold voltage, consequently degrading the device performance [6]. In ultra-scaled devices, the recovery phase is observed as sudden increments in the drain current, similar to RTN current shifts. In fact, there is a general consensus that RTN and the recoverable component of BTI are caused by the same type of defects [4],[7], and the observation of one phenomenon or the other actually depends on the operation conditions (voltage, time, temperature).

Then, nowadays both phenomena are usually described in the context of defect-centric models [7]-[10], which account for the variations in the transistor threshold voltage caused by the trapping/detrapping of charge carriers in/from defects present in the devices. To assess the impact on circuit performances it is crucial to characterize these phenomena and extract the statistical distributions of the main parameters that describe them.

The characterization of RTN and BTI defects traditionally relies on the detailed analysis of individual current traces [11]-[13]. This analysis detects the number of transitions with a distinct amplitude, each one of them corresponding to a different defect, or the total number of current levels. From any of these values, the number of defects can be easily calculated. However, these conventional approaches are convoluted and error-prone to the point that they are unable to correctly extract the number of defects and associated current shifts when: (1) the trace is very complex (i.e., with a large number of distinct transitions, particularly if some of these transitions have similar associated current amplitudes); (2) some of the defects produce current shifts with an amplitude below the noise level; or, (3) the defect characteristics are around the measurement

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equipment resolution. In this paper, we propose a method based on a new metric, the Maximum Current Fluctuation (*MCF*), to obtain the distribution of the number of active TDV-related defects (i.e., defects that suffer at least one trapping/detrapping event) and their associated current shifts, δI , that i) accounts for all active defects, and ii) does not require any complex analysis of the experimental current traces, since only a very rough estimation of the current shift distribution (easier to obtain) and the upper and lower bounds of the current are required. This model describes the experimental data accurately at gate voltages ranging from the near-threshold region to the nominal operation voltage of the technology.

The rest of this paper is structured as follows. First, in Section II, the *MCF* methodology used in this work is explained. Then, in Section III, the results obtained using this methodology are presented. Finally, in Section IV, conclusions are drawn.

II. METHODOLOGY DESCRIPTION

In the following, a metric is introduced to easily extract the defect-centric model parameters, i.e., the number of active defects and their associated current jumps, from experimental current traces.

Consider the current trace shown in Fig. 1a corresponding to a PMOS device of $W/L=80\text{nm}/60\text{nm}$ biased with $|V_{GS}| = 1.2\text{V}$ and $|V_{DS}| = 0.1\text{V}$. We define the cumulative maximum current (*CMAXC*) at any time instant t' within the experimental window as the maximum current within the interval $0 \leq t \leq t'$:

$$CMAXC(t') = \max_{t \in [0, t']} I(t) \quad (1)$$

CMAXC is plotted in the same figure with the orange line. Similarly, the cumulative minimum current (*CMINC*) at any time instant t' within the experimental window is defined as:

$$CMINC(t') = \min_{t \in [0, t']} I(t) \quad (2)$$

and is plotted in the same figure with a lower blue line. Finally, the metric proposed in this paper, the maximum current fluctuation (*MCF*), is defined as the difference between both:

$$MCF(t') = CMAXC(t') - CMINC(t') \quad (3)$$

Notice that, although current transitions caused by TDV defects are bipolar (i.e., charge trapping/de-trapping events decrease/increase the drain current), the *MCF* metric is, according to its definition, unipolar (i.e., the difference between the maximum and minimum values of the current will always be a positive value and *MCF*(t) an ever-increasing function).

Due to the TDV stochasticity, each device shows a different *MCF* trace. In fact, the *MCF* value for a given device at a time t can be expressed as:

$$MCF(t) = \sum_{i=1}^N \delta I_i + \varepsilon_{noise} \quad (4)$$

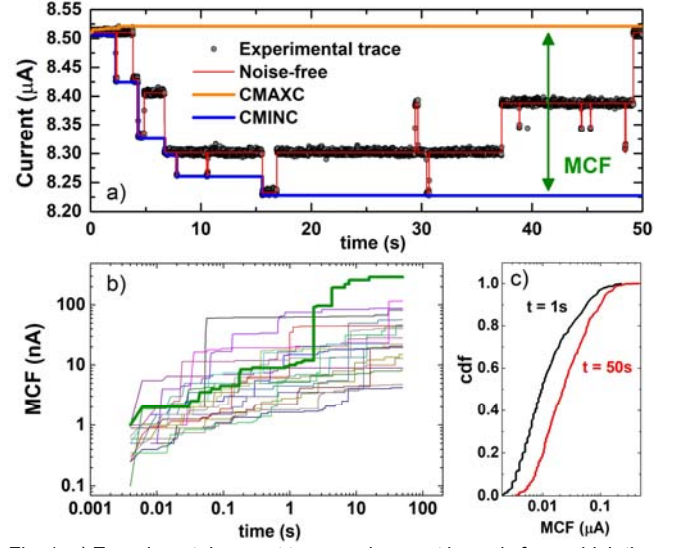


Fig. 1. a) Experimental current trace and current bounds from which the *MCF* is computed. The arrow indicates the *MCF* at around $t \approx 41\text{s}$. b) *MCF*(t) obtained from 20 current traces (the one in Fig. 1a is highlighted in green). c) The cumulative distribution functions (cdf) of the *MCF*s for $t=1\text{s}$ and $t=50\text{s}$ for a set of 500 transistors.

where N represents the number of active defects (i.e., defects that have captured/emitted a charge carrier at least once) from the beginning of the time window up to time t , and δI_i represents the current shift associated to the trapping/detrapping in/from each of those defects. The term ε_{noise} accounts for the contribution of the background noise to the *MCF*.

Fig. 1b shows the temporal evolution of the *MCF*s of 20 CMOS transistors. From these data, the cumulative distribution function (cdf) of the *MCF* can be obtained for a particular time instant, as shown in Fig. 1c. Notice how, in Fig. 1c, the distribution of the *MCF* values shift to larger values when longer times are considered. This is expected, since a larger number of active defects may be present in longer measurement windows (e.g., 50s) than in shorter ones (e.g., 1s). In fact, the *MCF* for a single device depends on the number of active defects, N , and their current shifts, δI , associated to their occupancy state, as shown in equation (4). Therefore, at each time t , there is a unique distribution of N that, together with the δI distribution, correctly describes the experimental *MCF* distribution. The extraction of both distributions is the goal of this work.

Note that the parameters (N and δI) that are needed by the *MCF*-based methodology to describe and predict the impact of TDV defects on the transistor characteristics, can be easily related to those of defect-centric TDV models, such as the Probabilistic Defect Occupancy (PDO) model [9]. The necessary parameters to construct such model are the total number of defects in the device (N_T), the distribution of the capture and emission time constants of the defects (τ_c , τ_e), which determine for each defect its probability of being occupied or empty, and the distribution of the amplitudes of the shifts associated to each defect when occupied, δI . These variables can be easily related to those used in our work. First, the number of active defects N extracted through the *MCF*-

based methodology depends on the total number of defects in the device, N_T , and on the time constants (τ_c , τ_e). For each defect, its time constants determine if the device will be active (i.e., suffer at least one trapping/detrapping event) during a given time window, in particular, during the time interval between the start of the measurement and the time instant in which the cdf of the *MCF* is calculated (e.g., a defect with $\tau_c \approx \tau_e \approx 1$ s will surely be accounted as an active defect during a measurement yielding the calculation of *MCF* at 50s). Finally, the distribution of the amplitudes of the shifts considered in defect-centric models is equivalent to the δI distribution considered in this work.

A related but different metric, the Within Device Fluctuation (*WDF*), has been presented in [14],[15]. However, though somehow related, they should not be confused, since they have been defined with different goals. The *WDF* metric for a time instant t' is given by the difference between $Max(t')$ and $Min(t')$. $Max(t')$ is given by the maximum current increase from the start of the measurement up to t' (hence, it is directly related to our definition of *CMAXC*), and $Min(t')$ is given by the minimum current variation from the end of the measurement window down to t' [14] (thus different from *CMINC*). A major difference between both metrics is that the *WDF* value at intermediate time instants depends on the size of the measurement time window, e.g., the value of $Min(t')$ for the trace in Fig. 1a for time instants between 18s and 30s is different if the measurement window extends to 30s or to 50s. This brings along an important consequence: the value of the *WDF* metric is only reliable for time instants sufficiently below the measurement window [14]. In any case, the *WDF* metric is, in itself, not appropriate for our purposes in this paper, since the $Min(t')$ values at intermediate time instants would be determined by the current variation at the later time instants. Hence, it cannot be related to the number of active defects and associated current jumps at those intermediate time instants, which is the goal of this paper.

To expose the limitations of time-consuming conventional approaches, an attempt has been made to extract the amplitudes of the current shifts caused by the trapping/de-trapping events of individual defects from experimental data. In order to get statistically relevant data, RTN traces have been measured in 500 PMOS transistors of $W/L = 80\text{nm}/60\text{nm}$ (integrated into an IC fabricated in a commercial 65nm planar CMOS technology specifically designed for TDV characterization) with $|V_{DS}| = 0.1\text{V}$ and different gate voltages, $|V_{GS}| = 0.6\text{V}, 0.7\text{V}, 0.8\text{V}, 1.0\text{V}$ and 1.2V , for 50s. Details about the IC and the measurement set-up can be found in [16] and [17]. Fig. 1a shows an example of noise-free RTN trace (red trace), obtained using the method in [11]. Noise-free trace here refers to the result of a mathematical processing that eliminates small current shifts due to background noise including e.g., thermal noise of the devices or noise of the measurement equipment, hence, obtaining a clean trace that clearly shows the effect of RTN. From the noise-free RTN traces, the δI caused by each defect is extracted. Using this procedure, hundreds of δI values were determined for each bias condition. Fig. 2a depicts the

experimental cdfs obtained for the δI distributions for the different biasing voltages, which can be correctly fitted considering two lognormal distributions. Fig. 2b displays the corresponding pdfs of the fitted two-lognormal model. The inset in Fig. 2b shows the experimental histogram constructed with the values of δI extracted at one of the bias conditions, which reinforces the suitability of a two lognormal model. The two lognormal function in the RTN amplitude distribution could originate from two categories of traps, located above the percolation path or not, which cause larger or smaller RTN amplitudes, respectively [18]. In any case, notice that no assumption has been done for the selection of this type of distribution, but rather has been found to properly fit the experimental data. In fact, only by using high resolution equipment, which allows the measurement of current jumps in the nanoampere range, the lognormal corresponding to the smaller amplitudes shows up.

Using these distributions and assuming a Poisson distribution with mean value $\langle N \rangle$ for the number of defects [19], the experimental *MCF* distributions can be fitted. An example is shown in Fig 3 (blue line) for $|V_{GS}| = 1.2\text{V}$ and $t =$

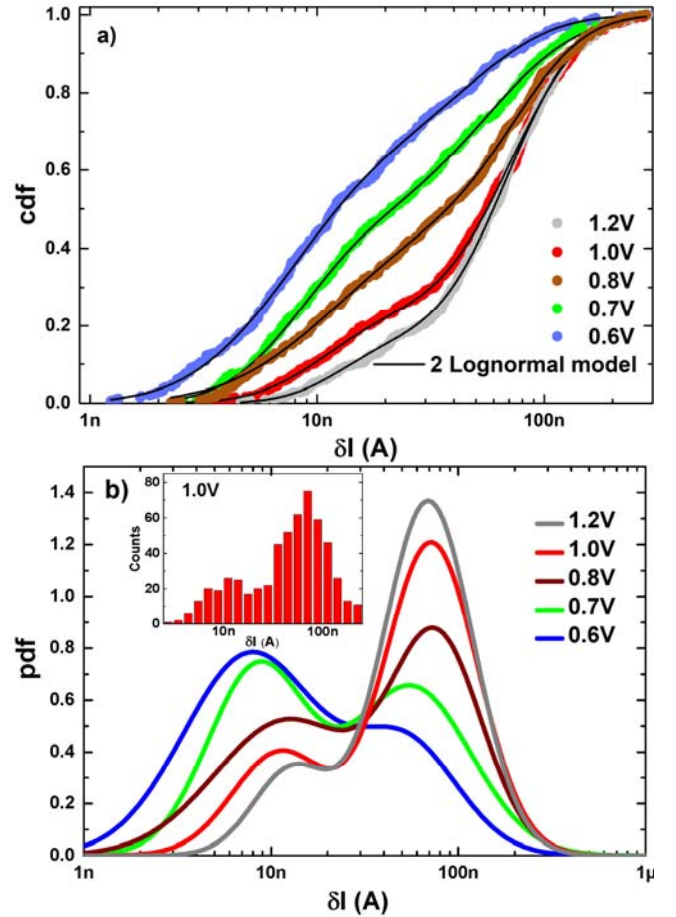


Fig. 2. a) δI cdf and fitted two-lognormal model (black lines). b) Corresponding probability density functions (pdf) of the fitted two-lognormal model. The inset in b) displays the histogram for the experimentally-extracted amplitudes for one of the bias conditions ($|V_{GS}| = 1.0\text{V}$).

100ms, which has made use of the 1.2V- δI distribution in Fig. 2. Clearly, the larger values (above 50nA) of the MCF distribution can be properly described using this approach. However, it fails to describe the smaller MCF values, because the δI distribution is obtained from the experimental current traces (as that in Fig. 1a), where, as mentioned in Section I, a portion of defects, especially those with small δI , have not been detected (due to experimental limitations) and yet these affect the MCF calculation, and are therefore accounted for by the MCF -based methodology. In fact, low- δI detection problem is a common issue of any method extracting δI , and consequently N , from experimental RTN traces [11], [19], [20], which are not able to detect defects that have an associated low-amplitude current shift. This reinforces the idea that conventional processing techniques may not only require a complex and time-consuming analysis, but may also lead to an erroneous characterization of the distribution of the number of defects and their associated amplitudes.

To solve this inaccuracy, we postulate here that the MCF metric can be useful even without a detailed, time-consuming analysis of the traces to attain a δI distribution (as in Fig. 2). In fact, these painstakingly-derived δI distributions are not really required. To illustrate this, the N distribution and a corrected δI distribution (red curve in the inset in Fig. 3) have been obtained by minimizing the difference between the MCF cdf in Fig. 3 (red curve) and the experimental one (grey dots). Notice that the δI distributions in Fig. 2 are not used for this MCF distributions fittings. Instead, for proper convergence of the fitting procedure, a very rough initial approximation (green curve in the inset in Fig. 3) is obtained from those traces (150 out of 500) that contain one clear defect (only two current levels separated by at least 5nA). This approximated δI distribution is faster and easier to extract and does not require a detailed, complex, and painstakingly processing of every shift, both large

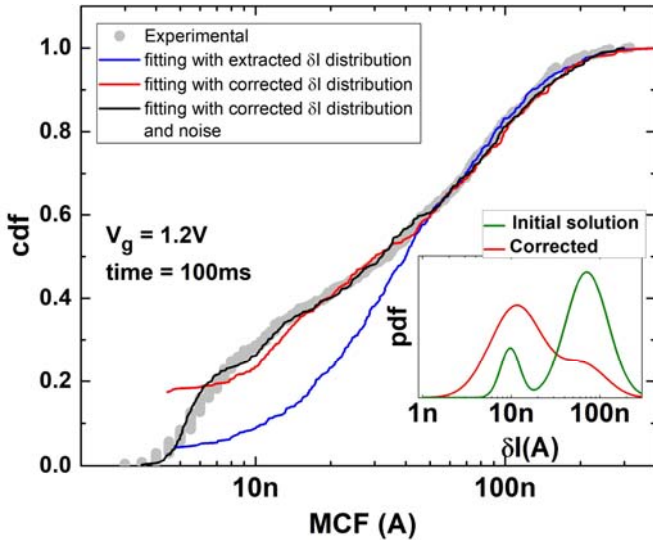


Fig. 3. MCF experimental distribution (grey symbols) and fitting (lines) considering the extracted and corrected δI distributions. The inset shows (green) the initial guess of the δI distribution considered for the MCF distribution fitting and (red) δI distribution that better fits the distribution.

and small, in every current trace, as required to get the distributions in Fig. 2.

An additional fact must be considered when trying to fit the experimental cdf of the MCF in Fig. 3: so far, devices without defects lead to simulated $MCF = 0$, while, experimentally, MCF will be always higher than zero due to the background noise. To solve this discrepancy, a Gaussian background noise with standard deviation obtained from the experimental data is added to the generation of the MCF values. Then, the experimental MCF distribution can be more accurately fitted (black line in Fig. 3) in the whole range. In summary, the experimental MCF distribution can be correctly described by just considering a Poisson distribution of N with mean value $\langle N \rangle$ (1.8, for the case shown in Fig. 3), the corrected δI distribution (red line in inset of Fig. 3) originated from an easy-to-attain, rough δI distribution approximation, and a Gaussian background noise.

III. RESULTS

Once the MCF -based methodology has been explained, the next step is to analyze the MCF time dependence. Fig. 4 shows the MCF distributions for different measurement intervals considering sections of the measured traces up to 10ms, 100ms, 1s, 10s and 50s. As expected, the MCF distributions shift to higher values for longer times. This is attributed to the fact that in longer current traces more defects can be activated, and, therefore, higher MCF values are encountered from the traces. Interestingly enough, the experimental MCF distributions can be well reproduced with our approach by simply varying the mean value of the active defects, $\langle N \rangle$, without any modification of the previously corrected δI distribution or the background noise. This implies that there is no correlation between the current shift associated to a defect and its time constants [21].

The approach illustrated above for $|V_{GS}| = 1.2V$ has been applied to other operation conditions. Fig. 5a shows the evolution of $\langle N \rangle$ with time for all the applied values of V_{GS} , which can be described using a bias-dependent lognormal-type distribution. Fig. 5b shows the Poisson distribution for the number of defects for $t=50s$ at different bias conditions.

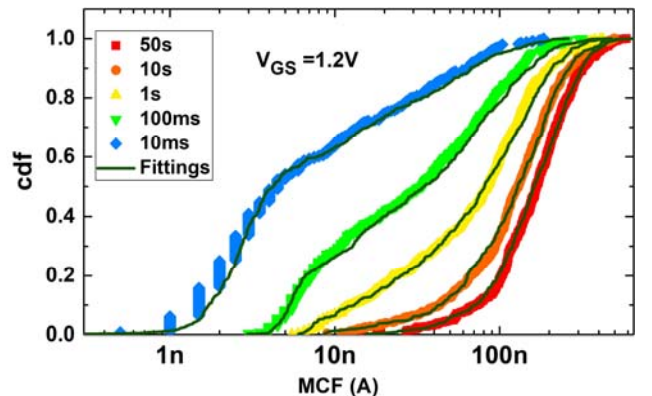


Fig. 4. Experimental MCF distributions obtained at different times (symbols) and fittings using the proposed technique (solid lines).

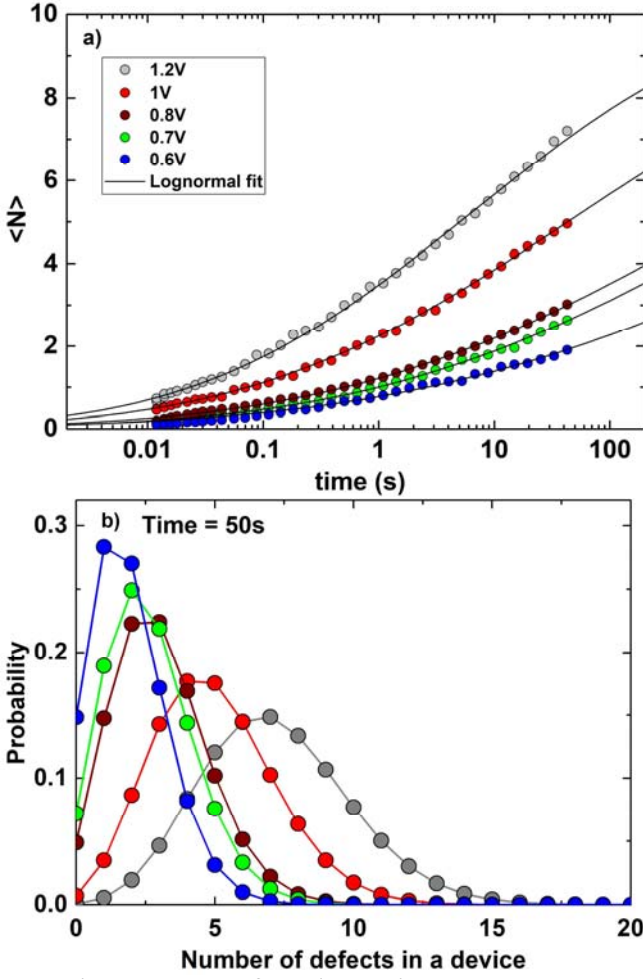


Fig. 5. a) Time evolution of $\langle N \rangle$ (symbols), which can be described using a lognormal-type distribution (lines). b) Poisson distributions at $t=50s$.

Table I displays the main parameters extracted through the MCF-based procedure, namely the distribution of the current shifts associated to the trapping/detrapping of defects, and the number of such active defects with respect to the operation time. Specifically, the parameters correspond to fitting the following distribution for the current shifts:

$$\text{cdf}(\delta I) = \frac{K}{2} \left[1 + \text{erf} \left(\frac{\log(\delta I) - \mu_l}{\sigma_l \sqrt{2}} \right) \right] + \frac{(1-K)}{2} \left[1 + \text{erf} \left(\frac{\log(\delta I) - \mu_u}{\sigma_u \sqrt{2}} \right) \right] \quad (5)$$

Where $\text{erf}()$ is the error function, μ_l , μ_u , σ_l and σ_u , represent the

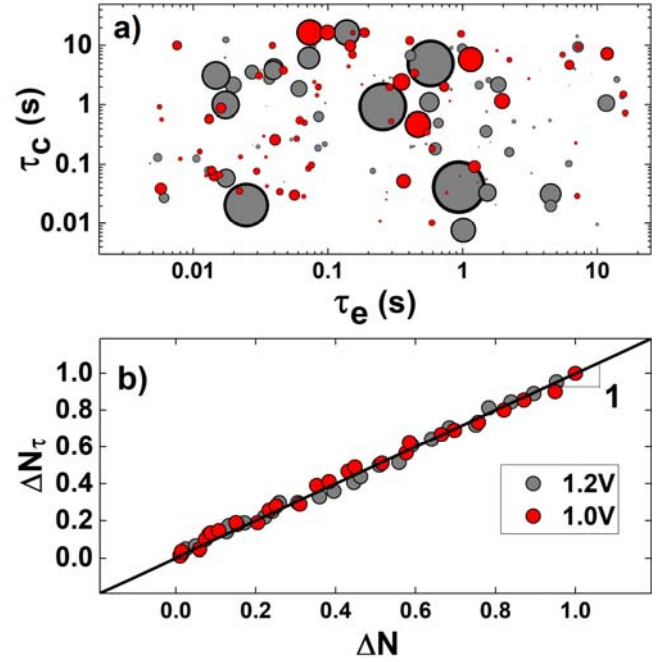


Fig. 6. a) Experimental τ_e and τ_c values obtained from the analysis of individual RTN traces. The size of the symbols represents the δI associated to each defect. b) Portion of the total active defects at a given time interval, ΔN , extracted using the MCF-based method, versus the relative number of defects, ΔN_t , obtained from the analysis of individual traces (Fig. 6a).

mean and standard deviation of the lower and upper lognormal function and K represents the relative amplitude of both distributions. And for the number of defects:

$$\langle N \rangle (t) = \frac{N_0}{2} \left[1 + \text{erf} \left(\frac{\log(t) - \mu_N}{\sigma_N \sqrt{2}} \right) \right] \quad (6)$$

Using these parameters, it is possible to predict the impact of the current fluctuations on the overall device drain current, as when using other defect centric models such as the Probabilistic Defect Occupancy (PDO) model [9].

Finally, to further illustrate the proposed methodology, the following test has been performed. From the detailed analysis of hundreds of individual RTN traces measured at $|V_{GS}| = 1V$ and $1.2V$, many emission (τ_e)/capture (τ_c) times have been extracted (Fig. 6a). With these data the number of defects N_t

TABLE I. PARAMETERS EXTRACTED FOR THE δI DISTRIBUTION OF RTN DEFECTS AND THE TIME EVOLUTION OF $\langle N \rangle$

$ V_{GS} $	K	μ_l	σ_l	μ_u	σ_u	N_0	μ_N	σ_N
0.6V	0.86	-7.95	0.07	-7.25	0.26	9.77	4.14	2.93
0.7V	0.77	-7.94	0.09	-7.16	0.22	9.87	3.21	2.51
0.8V	0.64	-8.05	0.13	-7.18	0.29	10.20	3.02	2.56
1V	0.76	-8.07	0.21	-7.15	0.25	9.85	1.59	2.14
1.2V	0.80	-7.91	0.12	-7.18	0.25	10.37	0.78	1.84

that will be active within a given time window has been calculated, assuming that defects will be active within a given time window t if their corresponding τ_e and τ_c are smaller than t . When the relative variation of the number of defects, ΔN_t , in a given time interval is plotted against the equivalent variation evaluated using the *MCF*-based method, ΔN , a linear relation between them for the voltages considered is observed (Fig. 6b). This result indicates that the same temporal trend is obtained for the number of defects that will be active, whether predicted by the novel *MCF*-based technique presented in this paper or extracted by performing a time-consuming, massive analysis of individual current traces. However, the procedure here presented is much simpler and more straightforward than the identification of individual defects from a large set of traces, while accounts for all the active defects in the device within the experimental time window, regardless of their amplitude.

IV. CONCLUSIONS

In deeply-scaled CMOS technologies, TDV phenomena, such as RTN and BTI, display a stochastic nature, caused by the trapping/de-trapping of charge carriers in/from individual defects. To account for this stochastic behavior, it is fundamental to develop accurate and automated methods to accurately characterize those phenomena.

This paper presents a novel methodology using a new metric, *MCF*, to evaluate the distributions of the number of active TDV-related defects and their associated current shifts within a given time window and under given operation conditions. Unlike conventional methods that analyze current traces, this method relies on a rather uncomplicated processing. It has also been shown that the *MCF*-based methodology is able to obtain a more precise distribution of the defect current amplitudes than conventional techniques, often failing at detecting defects with low-amplitude current shifts.

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